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(54) **SEMICONDUCTOR MEMORY DEVICE**

(71) Applicant: **Semiconductor Energy Laboratory Co., Ltd.**, Atsugi-shi, Kanagawa-ken (JP)

(72) Inventor: **Yasuhiko Takemura**, Kanagawa (JP)

(73) Assignee: **Semiconductor Energy Laboratory Co., Ltd.**, Atsugi-shi, Kanagawa-ken (JP)

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(56) **References Cited**

U.S. PATENT DOCUMENTS

4,777,625 A 10/1988 Sakui et al.
4,905,194 A 2/1990 Ohtsuka et al.

(Continued)

FOREIGN PATENT DOCUMENTS

EP 1737044 A 12/2006
EP 2120267 A 11/2009

(Continued)

OTHER PUBLICATIONS

Asakuma.N et al., "Crystallization and Reduction of Sol-Gel-Derived Zinc Oxide Films by Irradiation with Ultraviolet Lamp", Journal of Sol-Gel Science and Technology, 2003, vol. 26, pp. 181-184.

(Continued)

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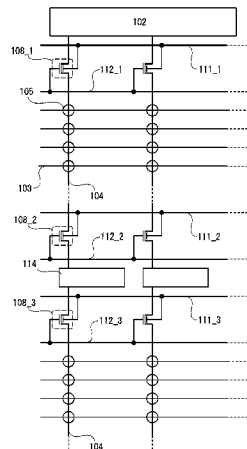
(74) *Attorney, Agent, or Firm* — Fish & Richardson P.C.

(57)

ABSTRACT

An object is to increase the retention characteristics of a memory device formed using a semiconductor with a wide bandgap, such as an oxide semiconductor. A transistor including a back gate (a back gate transistor) is inserted in series at one end of a bit line so that the back gate is constantly at a sufficiently negative potential. The minimum potential of the bit line is set higher than that of a word line. When power is turned off, the bit line is cut off by the back gate transistor, ensuring prevention of outflow of charge accumulated in the bit line. At this time, the potential of a source or a drain (bit line) of a cell transistor is sufficiently higher than that of a gate of the cell transistor (0 V), so that the cell transistor is put in a sufficiently off state; thus, data can be retained.

20 Claims, 10 Drawing Sheets



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- (56) **References Cited**
U.S. PATENT DOCUMENTS
- 5,148,393 A * 9/1992 Furuyama H01L 27/10805
257/E27.085
- 5,161,121 A 11/1992 Cho
5,305,263 A 4/1994 Morgan
5,414,662 A 5/1995 Foss et al.
5,610,533 A 3/1997 Arimoto et al.
5,731,856 A 3/1998 Kim et al.
5,744,864 A 4/1998 Cillessen et al.
6,294,274 B1 9/2001 Kawazoe et al.
6,477,098 B1 11/2002 Raad
6,563,174 B2 5/2003 Kawasaki et al.
6,707,720 B2 3/2004 Kamei et al.
6,727,522 B1 4/2004 Kawasaki et al.
6,771,531 B2 8/2004 Nishihara
6,781,915 B2 8/2004 Arimoto et al.
7,049,190 B2 5/2006 Takeda et al.
7,061,014 B2 6/2006 Hosono et al.
7,064,346 B2 6/2006 Kawasaki et al.
7,105,868 B2 9/2006 Nause et al.
7,211,825 B2 5/2007 Shih et al.
7,224,024 B2 * 5/2007 Forbes G11C 11/404
257/296
- 7,282,782 B2 10/2007 Hoffman et al.
7,297,977 B2 11/2007 Hoffman et al.
7,323,356 B2 1/2008 Hosono et al.
7,385,224 B2 6/2008 Ishii et al.
7,402,506 B2 7/2008 Levy et al.
7,411,209 B2 8/2008 Endo et al.
7,453,065 B2 11/2008 Saito et al.
7,453,087 B2 11/2008 Iwasaki
7,462,862 B2 12/2008 Hoffman et al.
7,468,304 B2 12/2008 Kaji et al.
7,501,293 B2 3/2009 Ito et al.
7,674,650 B2 3/2010 Akimoto et al.
7,732,819 B2 6/2010 Akimoto et al.
7,826,298 B2 11/2010 Nakai et al.
7,852,704 B2 12/2010 Takahashi et al.
8,014,221 B2 * 9/2011 Lee G11C 7/065
365/207
- 8,036,048 B2 * 10/2011 Takahashi G11C 8/08
365/149
- 8,067,775 B2 11/2011 Miyairi et al.
8,072,798 B2 12/2011 Takeyama
8,384,076 B2 2/2013 Park et al.
8,891,285 B2 * 11/2014 Takemura G11C 11/403
365/149
- 2001/0046027 A1 11/2001 Tai et al.
2002/0056838 A1 5/2002 Ogawa
2002/0121659 A1 9/2002 Hur
2002/0132454 A1 9/2002 Ohtsu et al.
2003/0189401 A1 10/2003 Kido et al.
2003/0218222 A1 11/2003 Wager, III et al.
2004/0038446 A1 2/2004 Takeda et al.
2004/0127038 A1 7/2004 Garcia et al.
2005/0017302 A1 1/2005 Hoffman
2005/0199959 A1 9/2005 Chiang et al.
2006/0035452 A1 2/2006 Garcia et al.
2006/0043377 A1 3/2006 Hoffman et al.
2006/0091793 A1 5/2006 Baude et al.
- 2006/0108529 A1 5/2006 Saito et al.
2006/0108636 A1 5/2006 Sano et al.
2006/0110867 A1 5/2006 Yabuta et al.
2006/0113536 A1 6/2006 Kumomi et al.
2006/0113539 A1 6/2006 Sano et al.
2006/0113549 A1 6/2006 Den et al.
2006/0113565 A1 6/2006 Abe et al.
2006/0169973 A1 8/2006 Isa et al.
2006/0170111 A1 8/2006 Isa et al.
2006/0197092 A1 9/2006 Hoffman et al.
2006/0208977 A1 9/2006 Kimura
2006/0228974 A1 10/2006 Thelss et al.
2006/0231882 A1 10/2006 Kim et al.
2006/0238135 A1 10/2006 Kimura
2006/0244107 A1 11/2006 Sugihara et al.
2006/0284171 A1 12/2006 Levy et al.
2006/0284172 A1 12/2006 Ishii
2006/0292777 A1 12/2006 Dunbar
2007/0024187 A1 2/2007 Shin et al.
2007/0046191 A1 3/2007 Saito
2007/0052025 A1 3/2007 Yabuta
2007/0054507 A1 3/2007 Kaji et al.
2007/0090365 A1 4/2007 Hayashi et al.
2007/0108446 A1 5/2007 Akimoto
2007/0152217 A1 7/2007 Lai et al.
2007/0172591 A1 7/2007 Seo et al.
2007/0187678 A1 8/2007 Hirao et al.
2007/0187760 A1 8/2007 Furuta et al.
2007/0194379 A1 8/2007 Hosono et al.
2007/0252928 A1 11/2007 Ito et al.
2007/0272922 A1 11/2007 Kim et al.
2007/0287296 A1 12/2007 Chang
2008/0006877 A1 1/2008 Mardilovich et al.
2008/0038882 A1 2/2008 Takechi et al.
2008/0038929 A1 2/2008 Chang
2008/0050595 A1 2/2008 Nakagawara et al.
2008/0073653 A1 3/2008 Iwasaki
2008/0083950 A1 4/2008 Pan et al.
2008/0106191 A1 5/2008 Kawase
2008/0128689 A1 6/2008 Lee et al.
2008/0129195 A1 6/2008 Ishizaki et al.
2008/0166834 A1 7/2008 Kim et al.
2008/0182358 A1 7/2008 Cowdery-Corvan et al.
2008/0224133 A1 9/2008 Park et al.
2008/0254569 A1 10/2008 Hoffman et al.
2008/0258139 A1 10/2008 Ito et al.
2008/0258140 A1 10/2008 Lee et al.
2008/0258141 A1 10/2008 Park et al.
2008/0258143 A1 10/2008 Kim et al.
2008/0296568 A1 12/2008 Ryu et al.
2009/0068773 A1 3/2009 Lai et al.
2009/0073325 A1 3/2009 Kuwabara et al.
2009/0073744 A1 3/2009 Takahashi et al.
2009/0114910 A1 5/2009 Chang
2009/0134399 A1 5/2009 Sakakura et al.
2009/0152506 A1 6/2009 Umeda et al.
2009/0152541 A1 6/2009 Maekawa et al.
2009/0251979 A1 10/2009 Chang
2009/0278122 A1 11/2009 Hosono et al.
2009/0280600 A1 11/2009 Hosono et al.
2010/0065844 A1 3/2010 Tokunaga
2010/0092800 A1 4/2010 Itagaki et al.
2010/0109002 A1 5/2010 Itagaki et al.
2010/0279474 A1 11/2010 Akimoto et al.
2011/0101351 A1 5/2011 Yamazaki
2011/0102409 A1 5/2011 Hayakawa
2011/0134680 A1 6/2011 Saito
2011/0156027 A1 6/2011 Yamazaki et al.
2011/0182110 A1 7/2011 Yamazaki et al.
2011/0199351 A1 8/2011 Kurokawa
2011/0317474 A1 12/2011 Kato
2012/0014157 A1 1/2012 Kato et al.
2012/0051118 A1 3/2012 Yamazaki et al.
2012/0061668 A1 3/2012 Miyairi et al.
- FOREIGN PATENT DOCUMENTS**
EP 2226847 A 9/2010
EP 2408011 A 1/2012

(56)

References Cited

FOREIGN PATENT DOCUMENTS

EP	2927965	A	10/2015
JP	60-198861	A	10/1985
JP	63-210022	A	8/1988
JP	63-210023	A	8/1988
JP	63-210024	A	8/1988
JP	63-215519	A	9/1988
JP	63-239117	A	10/1988
JP	63-265818	A	11/1988
JP	05-109274	A	4/1993
JP	05-251705	A	9/1993
JP	08-264794	A	10/1996
JP	09-245482	A	9/1997
JP	11-505377		5/1999
JP	11-328966	A	11/1999
JP	2000-044236	A	2/2000
JP	2000-150900	A	5/2000
JP	2002-076356	A	3/2002
JP	2002-289859	A	10/2002
JP	2003-086000	A	3/2003
JP	2003-086808	A	3/2003
JP	2003-217280	A	7/2003
JP	2004-103957	A	4/2004
JP	2004-273614	A	9/2004
JP	2004-273732	A	9/2004
JP	2009-278115	A	11/2009
WO	WO-2004/114391		12/2004

OTHER PUBLICATIONS

Asaoka.Y et al., "29.1: Polarizer-Free Reflective LCD Combined With Ultra Low-Power Driving Technology", SID Digest '09 : SID International Symposium Digest of Technical Papers, May 31, 2009, pp. 395-398.

Chern.H et al., "An Analytical Model for the Above-Threshold Characteristics of Polysilicon Thin-Film Transistors", IEEE Transactions on Electron Devices, Jul. 1, 1995, vol. 42, No. 7, pp. 1240-1246.

Cho.D et al., "21.2: Al and Sn-Doped Zinc Indium Oxide Thin Film Transistors for AMOLED Back-Plane", SID Digest '09 : SID International Symposium Digest of Technical Papers, May 31, 2009, pp. 280-283.

Clark.S et al., "First Principles Methods Using Castep", Zeitschrift fur Kristallographie, 2005, vol. 220, pp. 567-570.

Coates.D et al., "Optical Studies of the Amorphous Liquid-Cholesteric Liquid Crystal Transition: The "Blue Phase"", Physics Letters, Sep. 10, 1973, vol. 45A, No. 2, pp. 115-116.

Costello.M et al., "Electron Microscopy of a Cholesteric Liquid Crystal and Its Blue Phase", Phys. Rev. A (Physical Review. A), May 1, 1984, vol. 29, No. 5, pp. 2957-2959.

Dembo.H et al., "RFCPUS on Glass and Plastic Substrates Fabricated by TFT Transfer Technology", IEDM 05: Technical Digest of International Electron Devices Meeting, Dec. 5, 2005, pp. 1067-1069.

Fortunato.E et al., "Wide-Bandgap High-Mobility ZnO Thin-Film Transistors Produced at Room Temperature", Appl. Phys. Lett. (Applied Physics Letters), Sep. 27, 2004, vol. 85, No. 13, pp. 2541-2543.

Fung.T et al., "2-D Numerical Simulation of High Performance Amorphous In-Ga-Zn-O TFTs for Flat Pane Displays", AM-FPD '08 Digest of Technical Papers, Jul. 2, 2008, pp. 251-252, The Japan Society of Applied Physics.

Godo.H et al., "P-9: Numerical Analysis on Temperature Dependence of Characteristics of Amorphous In-Ga-Zn-Oxide TFT", SID Digest '09 : SID International Symposium Digest of Technical Papers, May 31, 2009, pp. 1110-1112.

Godo.H et al., "Temperature Dependence of Characteristics and Electronic Structure for Amorphous In-Ga-Zn-Oxide TFT", AM-FPD '09 Digest of Technical Papers, Jul. 1, 2009, pp. 41-44.

Hayashi.R et al., "42.1: Invited Paper: Improved Amorphous In-Ga-Zn-O TFTs", SID Digest '08 : SID International Symposium Digest of Technical Papers, May 20, 2008, vol. 39, pp. 621-624.

Hirao.T et al., "Novel Top-Gate Zinc Oxide Thin-Film Transistors (ZnO TFTs) for AMLCDs", J. Soc. Inf. Display (Journal of the Society for Information Display), 2007, vol. 15, No. 1, pp. 17-22.

Hosono.H et al., "Working hypothesis to explore novel wide band gap electrically conducting amorphous oxides and examples", J. Non-Cryst. Solids (Journal of Non-Crystalline Solids), 1996, vol. 198-200, pp. 165-169.

Hosono.H, "68.3: Invited Paper: Transparent Amorphous Oxide Semiconductors for High Performance TFT", SID Digest '07 : SID International Symposium Digest of Technical Papers, 2007, vol. 38, pp. 1830-1833.

Hsieh.H et al., "P-29: Modeling of Amorphous Oxide Semiconductor Thin Film Transistors and Subgap Density of States", SID Digest '08 : SID International Symposium Digest of Technical Papers, May 20, 2008, vol. 39, pp. 1277-1280.

Ikedo.T et al., "Full-Functional System Liquid Crystal Display Using CG-Silicon Technology", SID Digest '04 : SID International Symposium Digest of Technical Papers, 2004, vol. 35, pp. 860-863.

Janotti.A et al., "Native Point Defects in ZnO", Phys. Rev. B (Physical Review. B), Oct. 4, 2007, vol. 76, No. 16, pp. 165202-1-165202-22.

Janotti.A et al., "Oxygen Vacancies in ZnO", Appl. Phys. Lett. (Applied Physics Letters), 2005, vol. 87, pp. 122102-1-122102-3.

Jeong.J et al., "3.1: Distinguished Paper: 12.1-Inch WXGA AMOLED Display Driven by Indium-Gallium-Zinc Oxide TFTs Array", SID Digest '08 : SID International Symposium Digest of Technical Papers, May 20, 2008, vol. 39, No. 1, pp. 1-4.

Jin.D et al., "65.2: Distinguished Paper: World-Largest (6.5") Flexible Full Color Top Emission AMOLED Display on Plastic Film and its Bending Properties", SID Digest '09 : SID International Symposium Digest of Technical Papers, May 31, 2009, pp. 983-985.

Kanno.H et al., "White Stacked Electrophosphorescent Organic Light-Emitting Devices Employing MOO3 as a Charge-Generation Layer", Adv. Mater. (Advanced Materials), 2006, vol. 18, No. 3, pp. 339-342.

Kikuchi.H et al., "39.1: Invited Paper: Optically Isotropic Nano-Structured Liquid Crystal Composites for Display Applications", SID Digest '09 : SID International Symposium Digest of Technical Papers, May 31, 2009, pp. 578-581.

Kikuchi.H et al., "62.2: Invited Paper: Fast Electro-Optical Switching in Polymer-Stabilized Liquid Crystalline Blue Phases for Display Application", SID Digest '07 : SID International Symposium Digest of Technical Papers, 2007, vol. 38, pp. 1737-1740.

Kikuchi.H et al., "Polymer-Stabilized Liquid Crystal Blue Phases", Nature Materials, Sep. 2, 2002, vol. 1, pp. 64-68.

Kim.S et al., "High-Performance oxide thin film transistors passivated by various gas plasmas", 214th ECS Meeting, 2008, No. 2317, ECS.

Kimizuka.N et al., "Spinel, YBFE2O4, and YB2FE3O7 Types of Structures for Compounds in the In2O3 and SC2O3-A2O3-BO Systems [A: Fe, Ga or Al; B: Mg, Mn, Fe, Ni, Cu, or Zn] at Temperatures over 1000°C", Journal of Solid State Chemistry, 1985, vol. 60, pp. 382-384.

Kimizuka.N et al., "Syntheses and Single-Crystal Data of Homologous Compounds, In2O3(ZnO)m (m = 3, 4, and 5), InGaO3(ZnO)3, and Ga2O3(ZnO)m (m = 7, 8, 9, and 16) in the In2O3-ZnGa2O4-ZnO System", Journal of Solid State Chemistry, Apr. 1, 1995, vol. 116, No. 1, pp. 170-178.

Kitzerow.H et al., "Observation of Blue Phases in Chiral Networks", Liquid Crystals, 1993, vol. 14, No. 3, pp. 911-916.

Kurokawa.Y et al., "UHF Rfcus on Flexible and Glass Substrates for Secure RFID Systems", Journal of Solid-State Circuits, 2008, vol. 43, No. 1, pp. 292-299.

Lany.S et al., "Dopability, Intrinsic Conductivity, and Nonstoichiometry of Transparent Conducting Oxides", Phys. Rev. Lett. (Physical Review Letters), Jan. 26, 2007, vol. 98, pp. 045501-1-045501-4.

Lee.H et al., "Current Status of, Challenges to, and Perspective View of AM-OLED", IDW '06 : Proceedings of the 13th International Display Workshops, Dec. 7, 2006, pp. 663-666.

Lee.J et al., "World'S Largest (15-Inch) XGA AMLCD Panel Using IGZO Oxide TFT", SID Digest '08 : SID International Symposium Digest of Technical Papers, May 20, 2008, vol. 39, pp. 625-628.

(56)

References Cited

OTHER PUBLICATIONS

- Lee.M et al., "15.4:Excellent Performance of Indium-Oxide-Based Thin-Film Transistors by DC Sputtering", SID Digest '09 : SID International Symposium Digest of Technical Papers, May 31, 2009, pp. 191-193.
- Li.O et al., "Modulated Structures of Homologous Compounds InMO₃(ZnO)_m (M=In,Ga; m=Integer) Described by Four-Dimensional Superspace Group", Journal of Solid State Chemistry, 1998, vol. 139, pp. 347-355.
- Masuda.S et al., "Transparent thin film transistors using ZnO as an active channel layer and their electrical properties", J. Appl. Phys. (Journal of Applied Physics), Feb. 1, 2003, vol. 93, No. 3, pp. 1624-1630.
- Meiboom.S et al., "Theory of the Blue Phase of Cholesteric Liquid Crystals", Phys. Rev. Lett. (Physical Review Letters), May 4, 1981, vol. 46, No. 18, pp. 1216-1219.
- Miyasaka.M, "Sufla Flexible Microelectronics on Their Way to Business", SID Digest '07 : SID International Symposium Digest of Technical Papers, 2007, vol. 38, pp. 1673-1676.
- Mo.Y et al., "Amorphous Oxide TFT Backplanes for Large Size AMOLED Displays", IDW '08 : Proceedings of the 6th International Display Workshops, Dec. 3, 2008, pp. 581-584.
- Nakamura.M et al., "The phase relations in the In₂O₃-Ga₂ZnO₄-ZnO system at 1350°C", Journal of Solid State Chemistry, Aug. 1, 1991, vol. 93, No. 2, pp. 298-315.
- Nakamura.M, "Synthesis of Homologous Compound with New Long-Period Structure", NIRIM Newsletter, Mar. 1, 1995, vol. 150, pp. 1-4.
- Nomura.K et al., "Amorphous Oxide Semiconductors for High-Performance Flexible Thin-Film Transistors", Jpn. J. Appl. Phys. (Japanese Journal of Applied Physics), 2006, vol. 45, No. 5B, pp. 4303-4308.
- Nomura.K et al., "Carrier transport in transparent oxide semiconductor with intrinsic structural randomness probed using single-crystalline InGaO₃(ZnO)₅ films", Appl. Phys. Lett. (Applied Physics Letters), Sep. 13, 2004, vol. 85, No. 11, pp. 1993-1995.
- Nomura.K et al., "Room-Temperature Fabrication of Transparent Flexible Thin-Film Transistors Using Amorphous Oxide Semiconductors", Nature, Nov. 25, 2004, vol. 432, pp. 488-492.
- Nomura.K et al., "Thin-Film Transistor Fabricated in Single-Crystalline Transparent Oxide Semiconductor", Science, May 23, 2003, vol. 300, No. 5623, pp. 1269-1272.
- Nowatari.H et al., "60.2: Intermediate Connector With Suppressed Voltage Loss for White Tandem OLEDs", SID Digest '09 : SID International Symposium Digest of Technical Papers, May 31, 2009, vol. 40, pp. 899-902.
- Oba.F et al., "Defect energetics in ZnO: A hybrid Hartree-Fock density functional study", Phys. Rev. B (Physical Review B), 2008, vol. 77, pp. 245202-1-245202-6.
- Oh.M et al., "Improving the Gate Stability of ZnO Thin-Film Transistors With Aluminum Oxide Dielectric Layers", J. Electrochem. Soc. (Journal of the Electrochemical Society), 2008, vol. 155, No. 12, pp. H1009-H1014.
- Ohara.H et al., "21.3:4.0 in. QVGA AMOLED Display Using In-Ga-Zn-Oxide TFTs With a Novel Passivation Layer", SID Digest '09 : SID International Symposium Digest of Technical Papers, May 31, 2009, pp. 284-287.
- Ohara.H et al., "Amorphous In-Ga-Zn-Oxide TFTs with Suppressed Variation for 4.0 inch QVGA AMOLED Display", AM-FPD '09 Digest of Technical Papers, Jul. 1, 2009, pp. 227-230, The Japan Society of Applied Physics.
- Orita.M et al., "Amorphous transparent conductive oxide InGaO₃(ZnO)_m (m<4):a Zn₄s conductor", Philosophical Magazine, 2001, vol. 81, No. 5, pp. 501-515.
- Orita.M et al., "Mechanism of Electrical Conductivity of Transparent InGaZnO₄", Phys. Rev. B (Physical Review B), Jan. 15, 2000, vol. 61, No. 3, pp. 1811-1816.
- Osada.T et al., "15.2: Development of Driver-Integrated Panel using Amorphous In-Ga-Zn-Oxide TFT", SID Digest '09 : SID International Symposium Digest of Technical Papers, May 31, 2009, pp. 184-187.
- Osada.T et al., "Development of Driver-Integrated Panel Using Amorphous In-Ga-Zn-Oxide TFT", AM-FPD '09 Digest of Technical Papers, Jul. 1, 2009, pp. 33-36.
- Park.J et al., "Amorphous Indium-Gallium-Zinc Oxide TFTs and Their Application for Large Size AMOLED", AM-FPD '08 Digest of Technical Papers, Jul. 2, 2008, pp. 275-278.
- Park.J et al., "Dry etching of ZnO films and plasma-induced damage to optical properties", J. Vac. Sci. Technol. B (Journal of Vacuum Science & Technology B), Mar. 1, 2003, vol. 21, No. 2, pp. 800-803.
- Park.J et al., "Electronic Transport Properties of Amorphous Indium-Gallium-Zinc Oxide Semiconductor upon Exposure to Water", Appl. Phys. Lett. (Applied Physics Letters), 2008, vol. 92, pp. 072104-1-072104-3.
- Park.J et al., "High performance amorphous oxide thin film transistors with self-aligned top-gate structure", IEDM 09: Technical Digest of International Electron Devices Meeting, Dec. 7, 2009, pp. 191-194.
- Park.J et al., "Improvements in the Device Characteristics of Amorphous Indium Gallium Zinc Oxide Thin-Film Transistors by Ar Plasma Treatment", Appl. Phys. Lett. (Applied Physics Letters), Jun. 26, 2007, vol. 90, No. 26, pp. 262106-1-262106-3.
- Park.S et al., "Challenge to Future Displays: Transparent AMOLED Driven by PEALD Grown ZnO TFT", IMID '07 Digest, 2007, pp. 1249-1252.
- Park.Sang-Hee et al., "42.3: Transparent ZnO Thin Film Transistor for the Application of High Aperture Ratio Bottom Emission AM-OLED Display", SID Digest '08 : SID International Symposium Digest of Technical Papers, May 20, 2008, vol. 39, pp. 629-632.
- Prins.M et al., "A Ferroelectric Transparent Thin-Film Transistor", Appl. Phys. Lett. (Applied Physics Letters), Jun. 17, 1996, vol. 68, No. 25, pp. 3650-3652.
- Sakata.J et al., "Development of 4.0-in. AMOLED Display With Driver Circuit Using Amorphous In-Ga-Zn-Oxide TFTs", IDW '09 : Proceedings of the 16th International Display Workshops, 2009, pp. 689-692.
- Son.K et al., "42.4L: Late-News Paper: 4 Inch QVGA AMOLED Driven by the Threshold Voltage Controlled Amorphous GIZO (Ga₂O₃-In₂O₃-ZnO) TFT", SID Digest '08 : SID International Symposium Digest of Technical Papers, May 20, 2008, vol. 39, pp. 633-636.
- Takahashi.M et al., "Theoretical Analysis of IGZO Transparent Amorphous Oxide Semiconductor", IDW '08 : Proceedings of the 15th International Display Workshops, Dec. 3, 2008, pp. 1637-1640.
- Tsuda.K et al., "Ultra Low Power Consumption Technologies for Mobile TFT-LCDs", IDW '02 : Proceedings of the 9th International Display Workshops, Dec. 4, 2002, pp. 295-298.
- Ueno.K et al., "Field-Effect Transistor on SrTiO₃ With Sputtered Al₂O₃ Gate Insulator", Appl. Phys. Lett. (Applied Physics Letters), Sep. 1, 2003, vol. 83, No. 9, pp. 1755-1757.
- Van de Walle.C, "Hydrogen as a Cause of Doping in Zinc Oxide", Phys. Rev. Lett. (Physical Review Letters), Jul. 31, 2000, vol. 85, No. 5, pp. 1012-1015.
- Prince.B, Semiconductor Memories, Feb. 1, 1995, p. 158, Wiley.
- Taiwanese Office Action (Application No. 101120722) Dated Dec. 17, 2015.

* cited by examiner

FIG. 1A

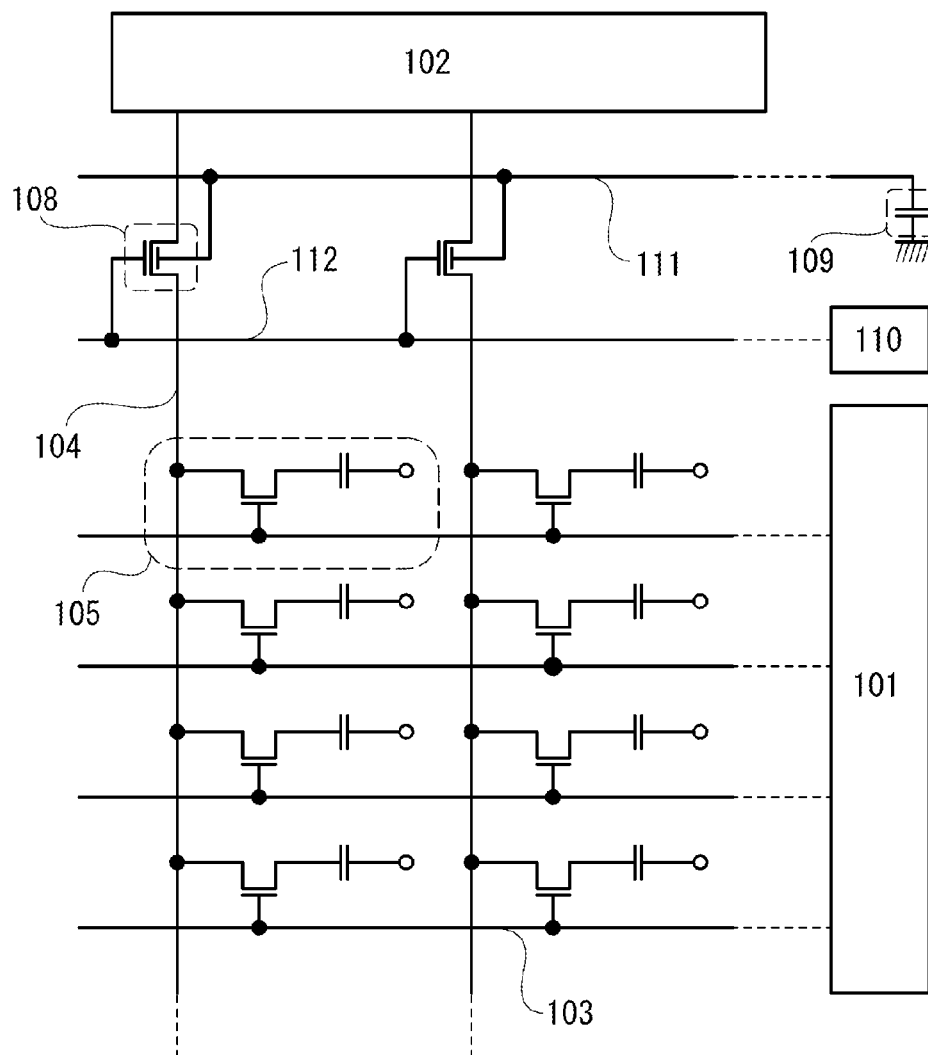


FIG. 1B

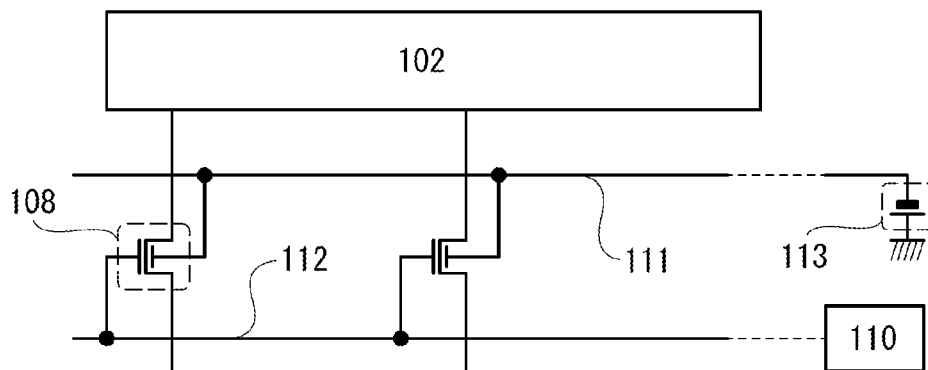


FIG. 2A

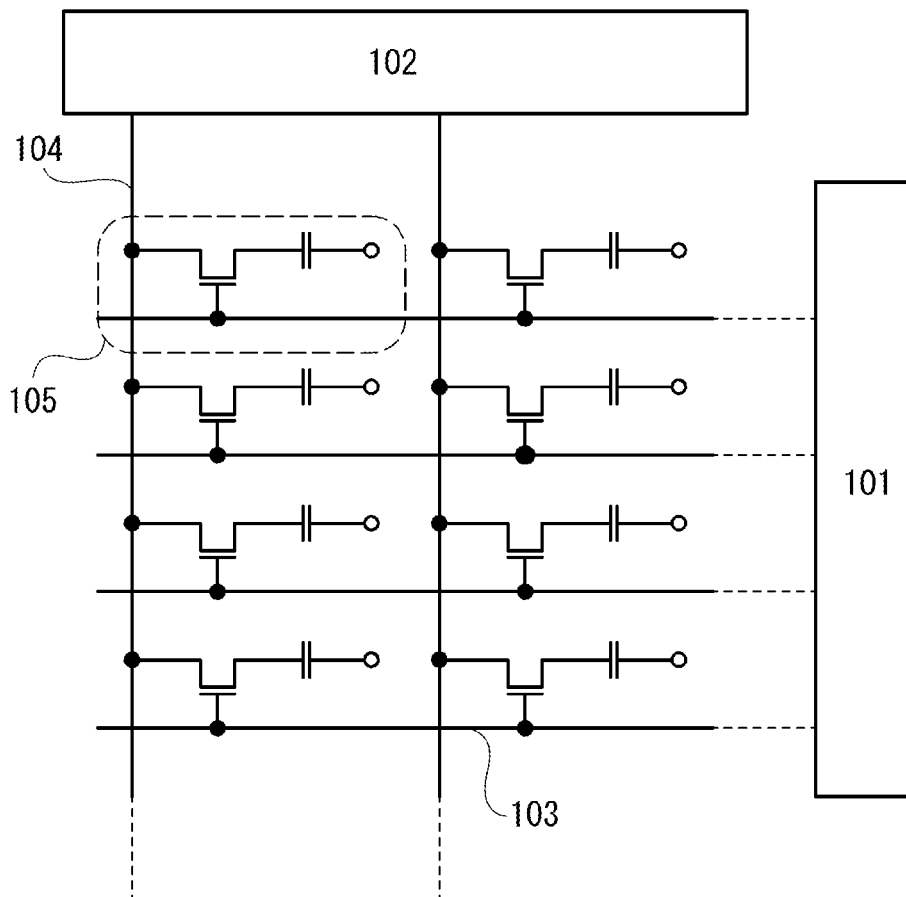


FIG. 2B

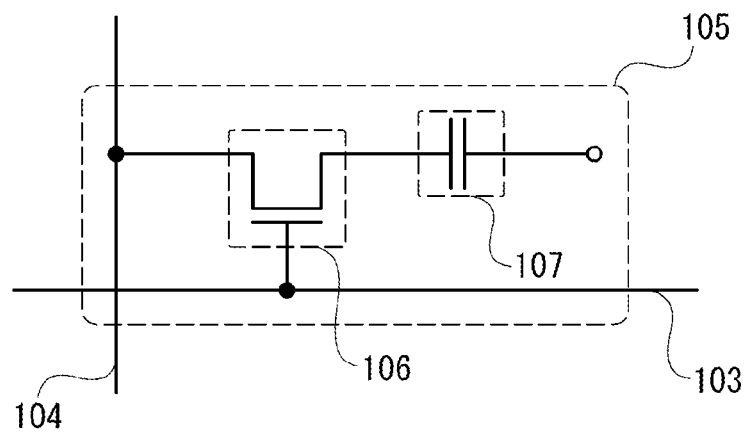


FIG. 3

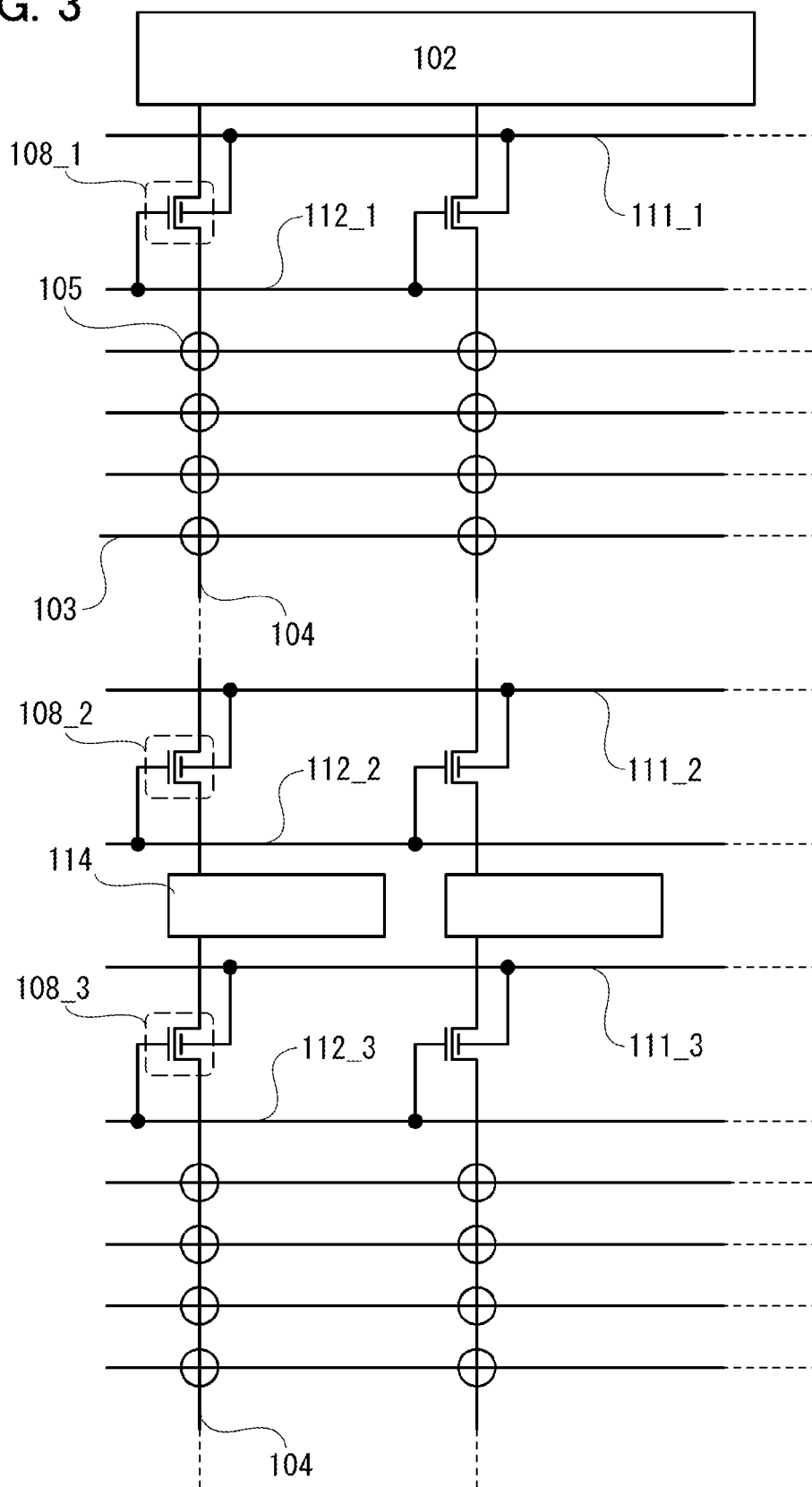


FIG. 4A

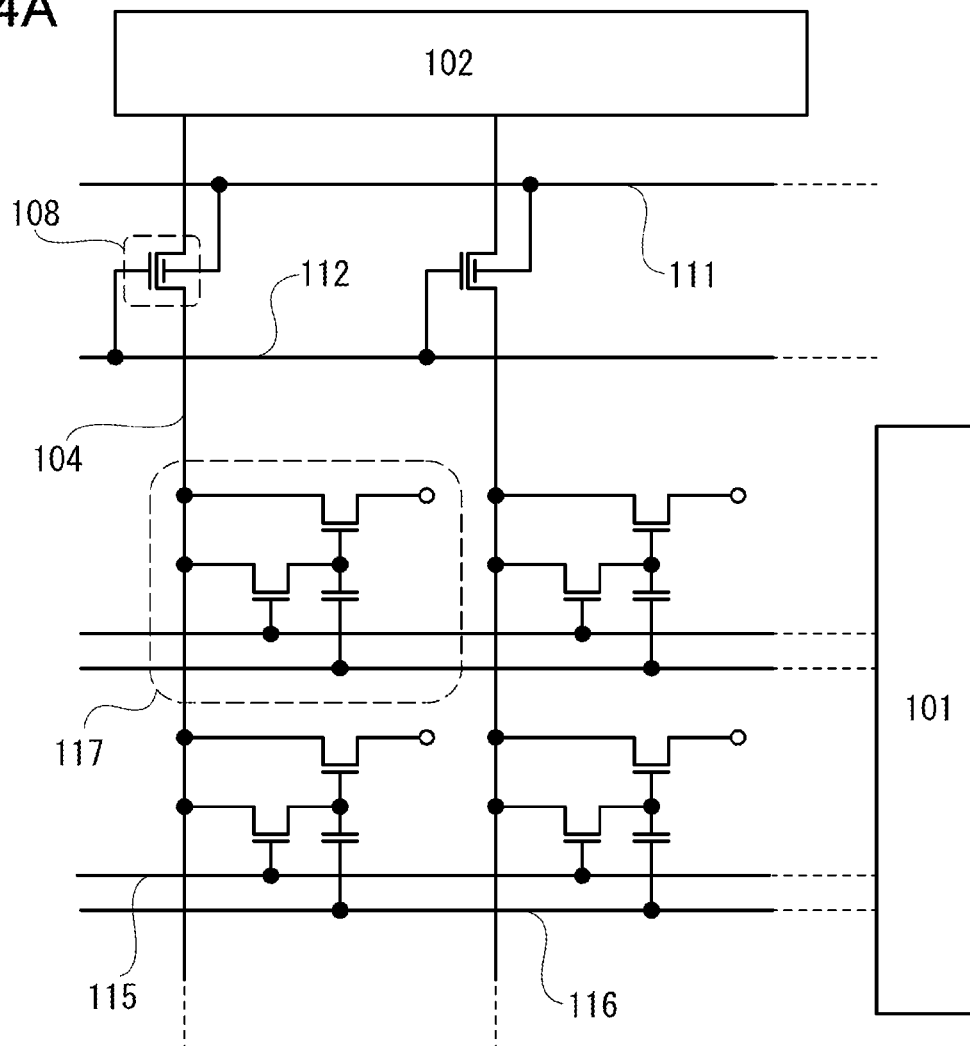


FIG. 4B

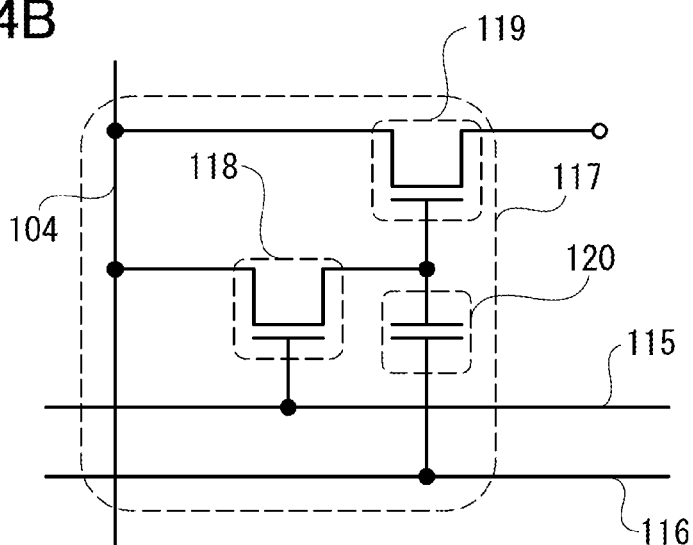


FIG. 5A

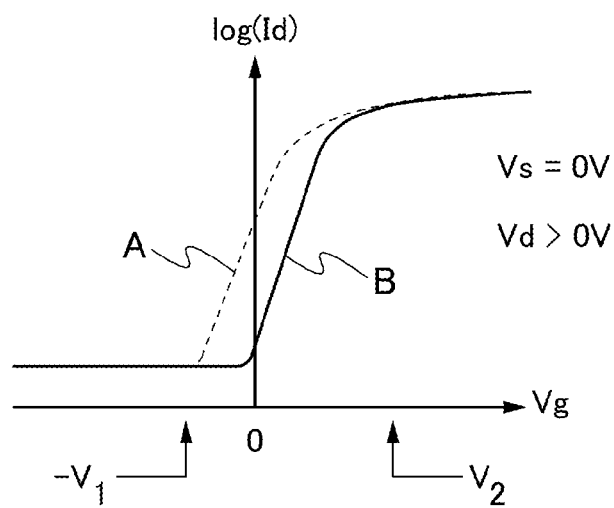


FIG. 5B

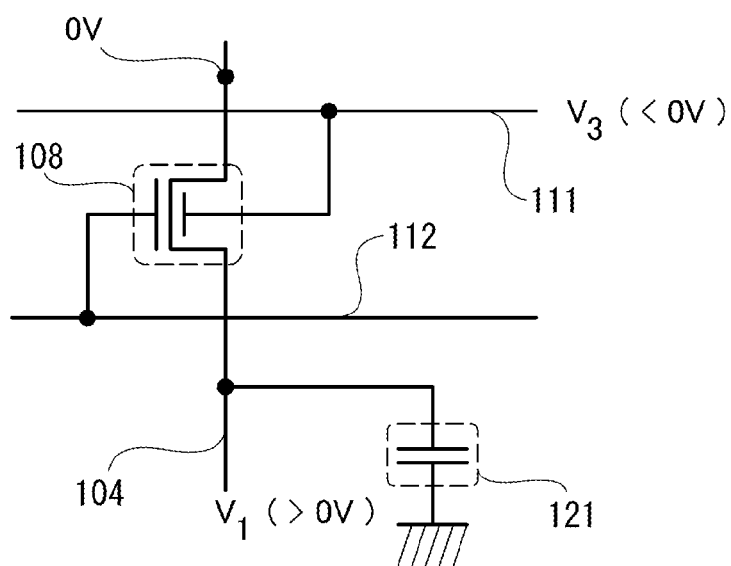


FIG. 5C

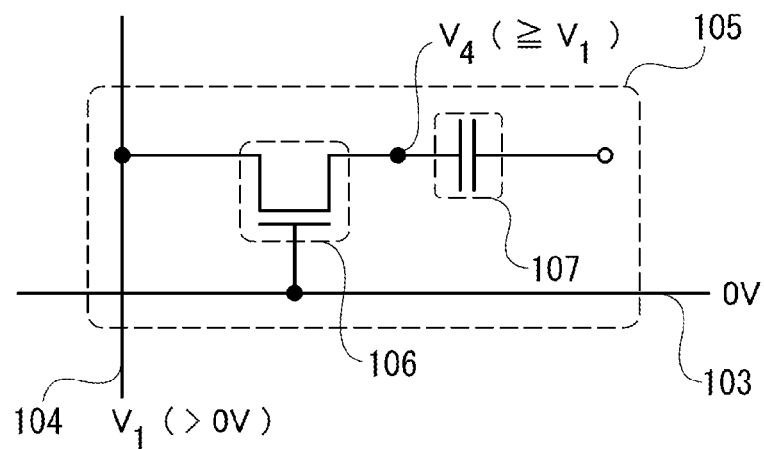


FIG. 6A

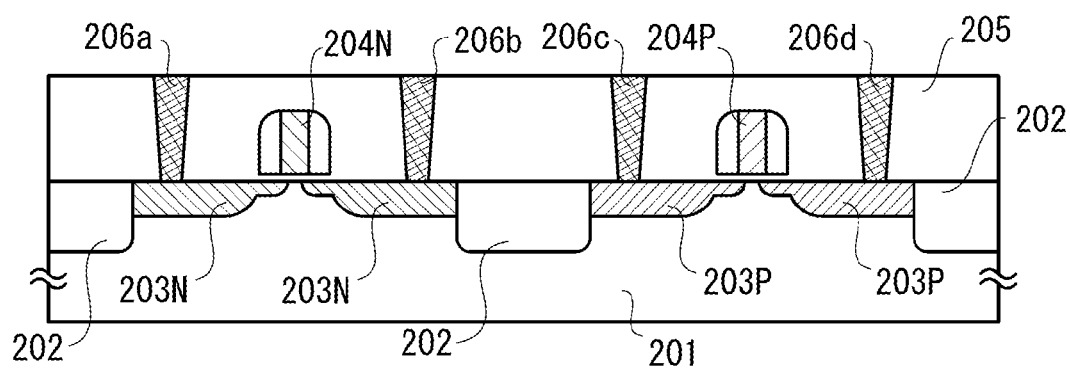


FIG. 6B

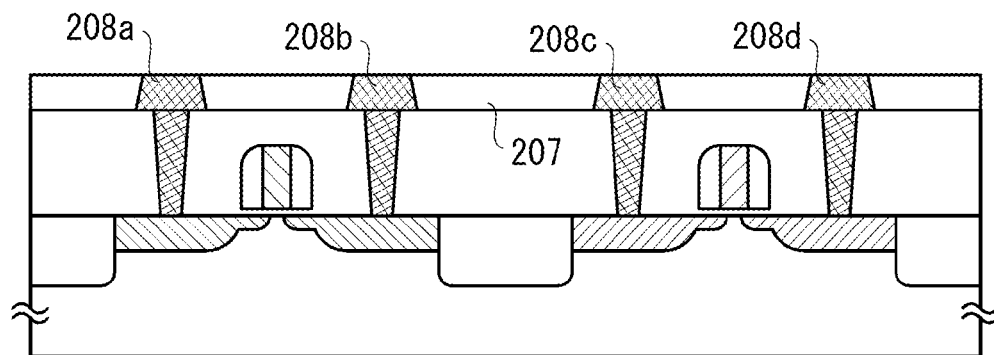


FIG. 6C

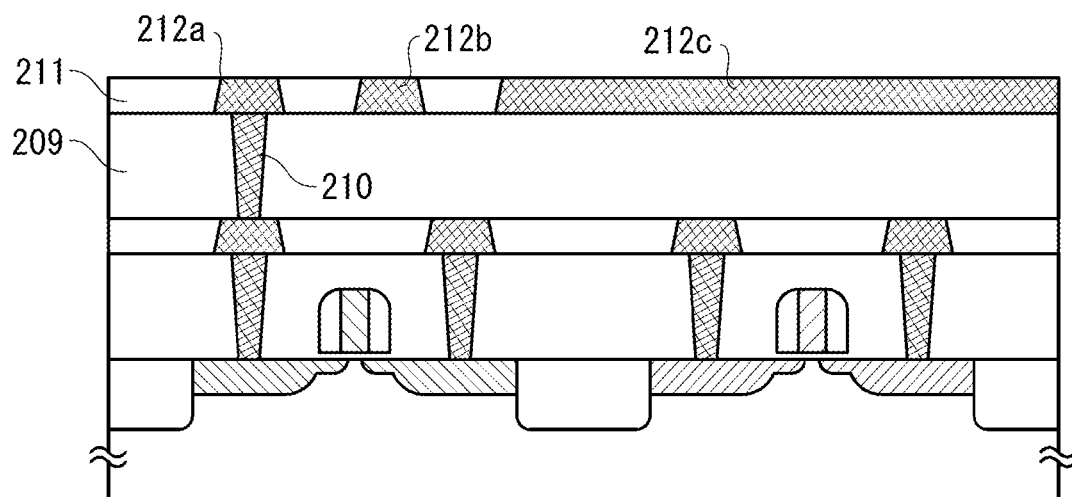


FIG. 7A

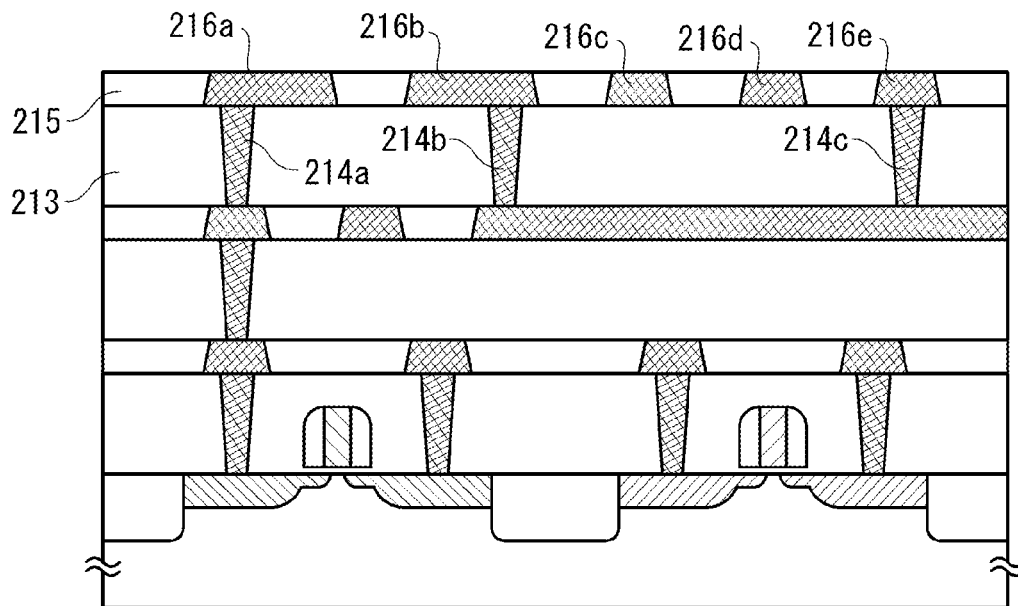


FIG. 7B

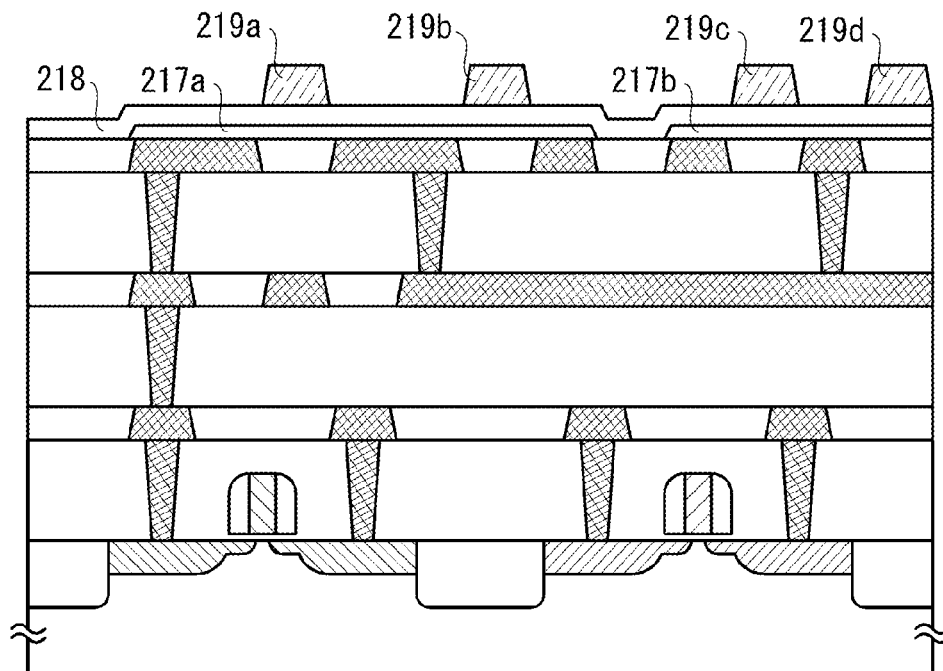


FIG. 8

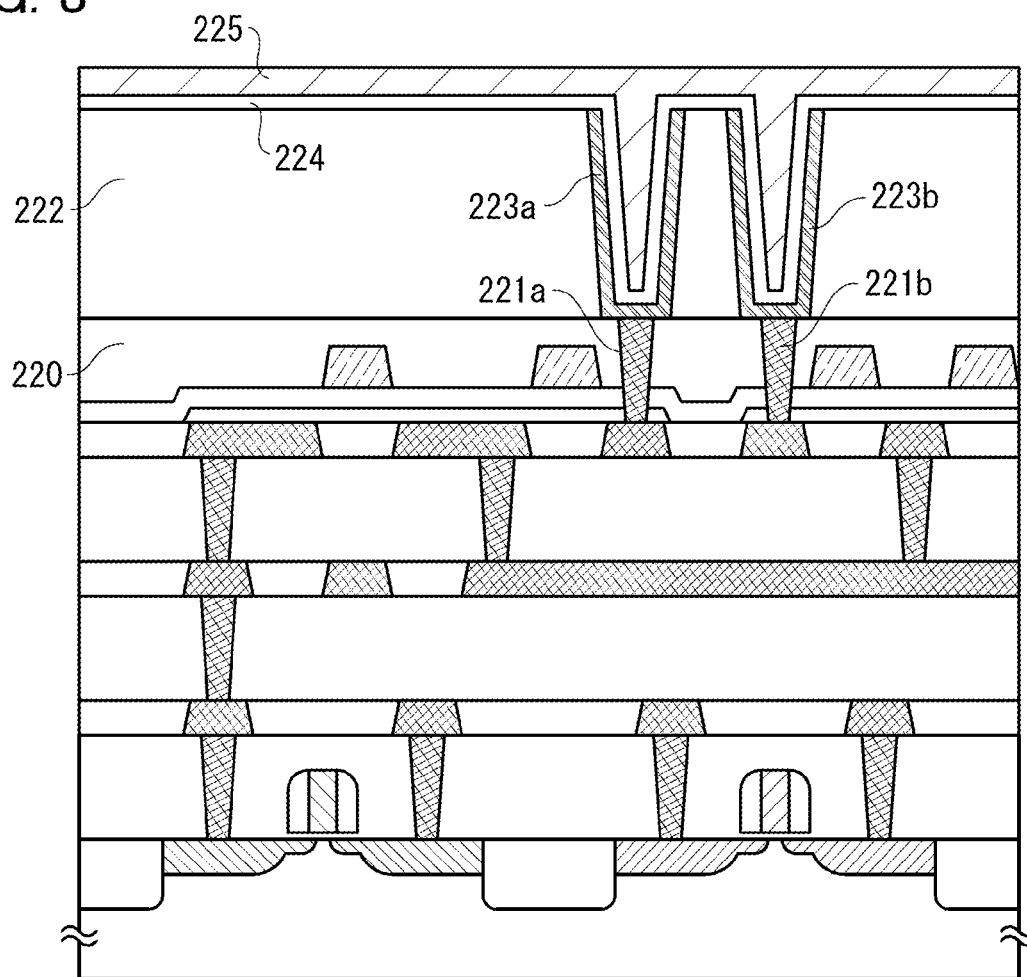


FIG. 9A

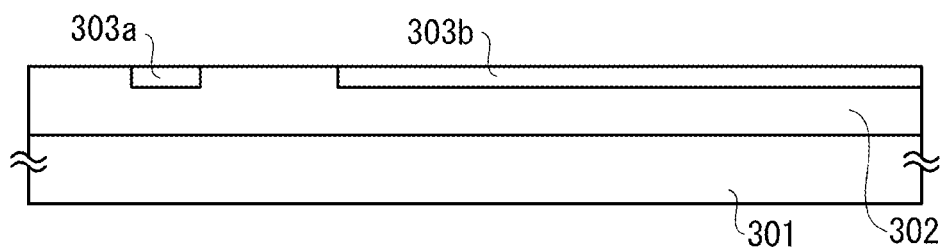


FIG. 9B

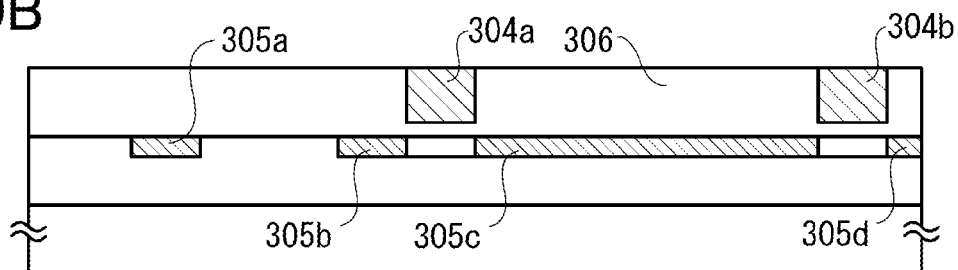


FIG. 9C

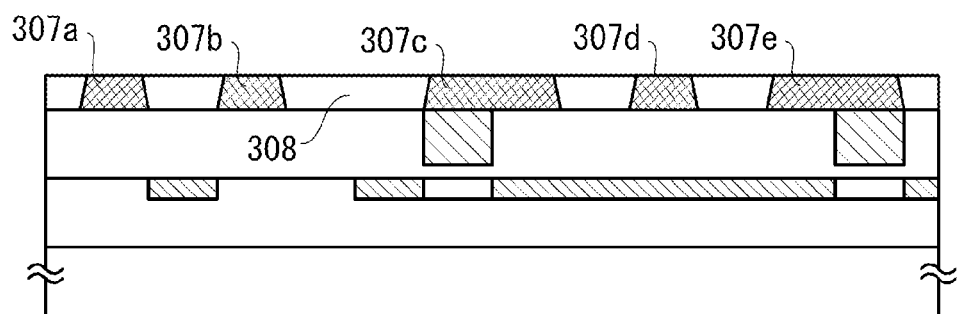


FIG. 9D

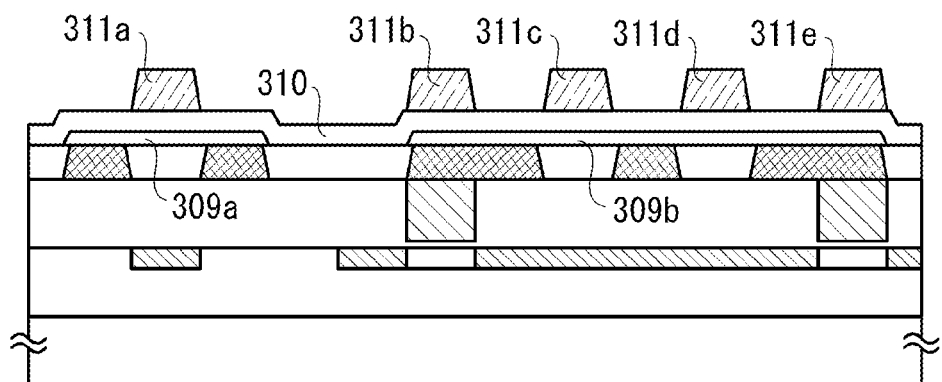


FIG. 10A

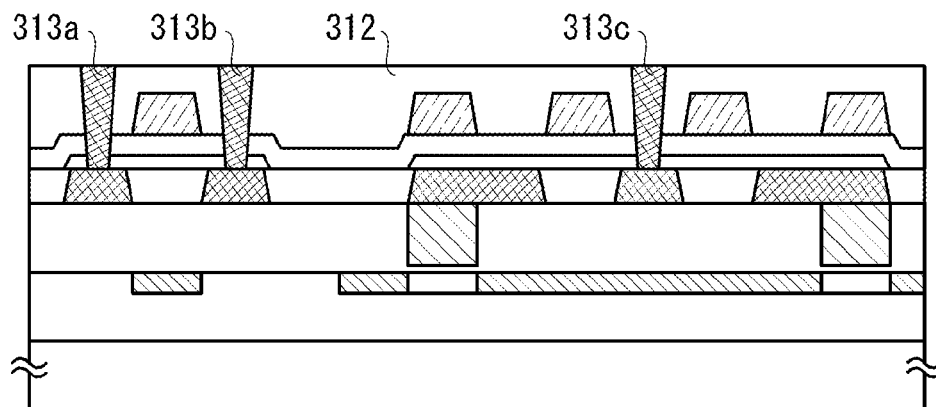


FIG. 10B

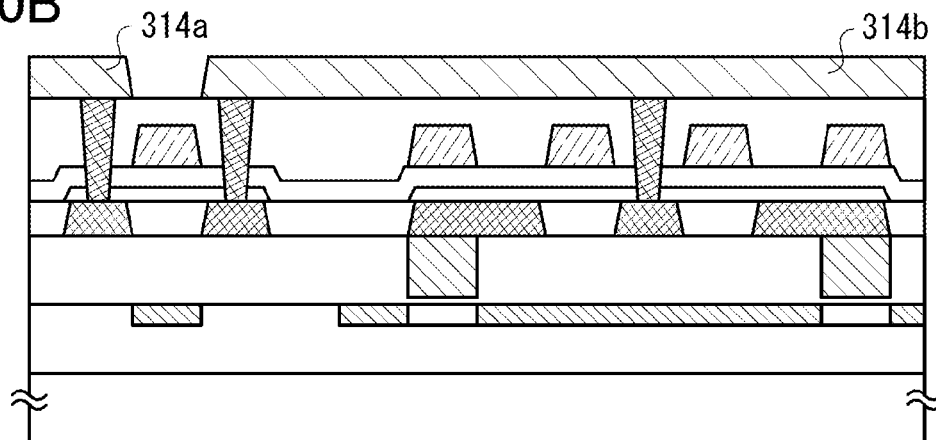
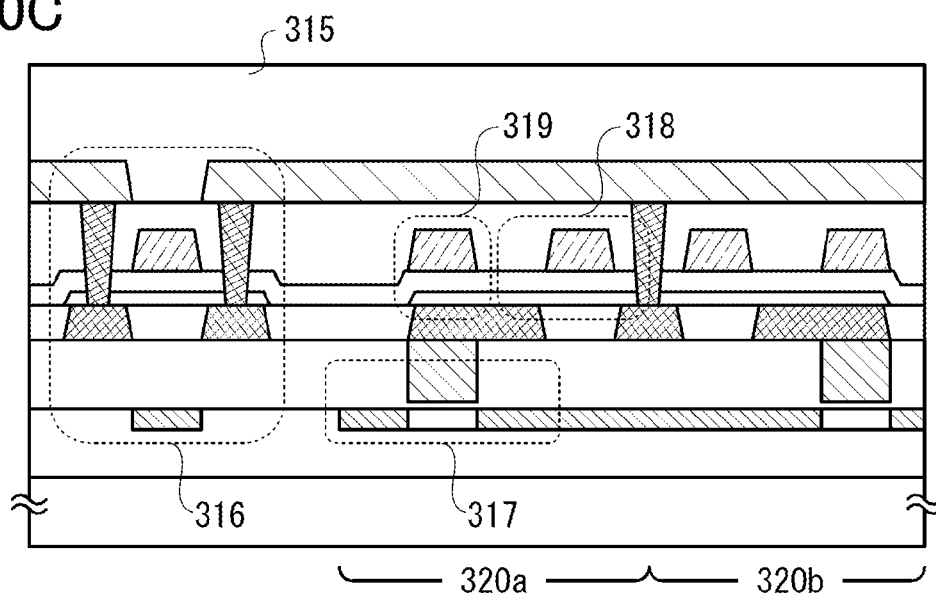


FIG. 10C



SEMICONDUCTOR MEMORY DEVICE

CROSS-REFERENCE TO RELATED APPLICATIONS

This application is a continuation of U.S. application Ser. No. 13/478,215, filed May 23, 2012, now U.S. Pat. No. 8,891,285, which claims the benefit of a foreign priority application filed in Japan as Serial No. 2011-129685 on Jun. 10, 2011, both of which are incorporated by reference.

BACKGROUND OF THE INVENTION

1. Field of the Invention

The present invention relates to a memory device formed using a semiconductor.

2. Description of the Related Art

Semiconductor memory devices include dynamic random access memories (DRAMs) (see Patent Document 1, for example). In a DRAM, memory cells **105** each including a cell transistor **106** and a capacitor **107** as illustrated in FIG. 2B are arranged in a matrix as illustrated in FIG. 2A, and a gate and a drain of the cell transistor **106** are connected to a word line **103** and a bit line **104**, respectively. Moreover, the DRAM includes a row driver **101** for driving a plurality of word lines and a column driver **102** for driving a plurality of bit lines.

The DRAM is powered by an external power supply to drive the row driver **101** and the column driver **102**. Note that a cell transistor formed using silicon semiconductor has a small drain current (off-state current) even in the off state, and thus requires tens of refresh operations (operations for replenishing the capacitor with charge) per second. In other words, the DRAM needs to be powered by the external power supply to retain a stored state.

In recent years, it has been found that charge can be retained for a very long period of time by utilizing the very low off-state current of transistors formed using an oxide semiconductor whose bandgap is two or more times that of silicon semiconductor. For example, the theoretical off-state current (drain current in the off state) of a semiconductor with a bandgap of 2.5 electron volts or more is 10^{-26} A or less. The use of a memory circuit utilizing this as a non-volatile memory has been proposed (see Patent Documents 2 to 4).

A transistor used in such a memory needs to exhibit sufficiently high off resistance (the resistance of the transistor in the off state), i.e., sufficiently low off-state current. For example, in order to retain charge in a capacitor of 30 fF, which is the capacitance of capacitors used in a DRAM in common use, for 10 years, a transistor exhibiting a resistance of as high as $1 \times 10^{22} \Omega$ or more in the off state is required. Assuming that the drain voltage is +1 V, the off-state current of the transistor needs to be 100 yA (1×10^{-22} A) or less.

The drain current of a transistor formed using an oxide semiconductor with a wide bandgap in the subthreshold region can be roughly estimated from the subthreshold value and the threshold voltage. The theoretical lower limit of the subthreshold value at room temperature (27° C.) is 60 mV/decade.

For example, assuming that the threshold voltage is +1 V, the subthreshold value is 60 mV/decade, and the drain current obtained when the threshold voltage is +1 V is 1 μ A (the source potential V_s is 0 V, while the drain potential V_d is +1 V), the drain current is 100 yA with a gate potential V_g of +40 mV. With a gate potential V_g of 0 V, the drain current

of the transistor is less than 100 yA, so that the charge in the capacitor can be retained for 10 years.

Note that the retention period is not limited to 10 years, and may be determined in the range from 10 seconds to 100 years depending on intended use. The capacitance of the capacitor or the off resistance or off-state current of the transistor may be set according to the retention period.

The above-described drain current is obtained at room temperature. In practice, some problems arise here. The subthreshold value depends on temperature. As temperature increases, the subthreshold value increases. Because it is also possible that the semiconductor memory device is stored at a high temperature, sufficient retention characteristics need to be also ensured at a temperature exceeding room temperature.

For example, the theoretical lower limit of the subthreshold value at 95° C. is 74 mV/decade. When the subthreshold value is 74 mV/decade, gate potential V_g with which the drain current becomes 100 yA is -180 mV. When the gate potential V_g is 0 V, the drain current is 10 zA (1×10^{-20} A), so that charge retention time is 1% of that at room temperature.

As transistor size is decreased, the subthreshold value increases owing to short channel effects. The conductivity type of silicon semiconductor can be controlled by doping. Therefore, in the case of an n-channel transistor, for example, short channel effects can be reduced by increasing the concentration of a p-type dopant in the channel formation region.

In contrast, the conductivity type of an oxide semiconductor cannot be controlled by controlling dopant concentration as in the case of silicon semiconductor. The intensity of one conductivity type of an oxide semiconductor can be changed, but the conductivity type of an oxide semiconductor cannot be reversed; for example, an n-type oxide semiconductor cannot be turned into a p-type one by doping. For this reason, short channel effects cannot be reduced by reversing the conductivity type of the channel formation region.

Therefore, with a channel length of 100 nm or less, the subthreshold value is 100 mV/decade or more, and the gate potential V_g needs to be maintained at -0.6 V or less. The threshold voltage is +1 V in the above description; when the threshold voltage is low, even the gate potential V_g at room temperature or with a long channel needs to be less than 0 V in order to sufficiently increase the off resistance. Note that the threshold voltage is dependent on the work function of a material for the gate; thus, it is difficult to increase the threshold voltage to +1.5 V or higher.

Under such conditions, data loss may occur when power from the external power supply to the semiconductor memory device is interrupted and the potential of the gate becomes the same as that of the source (i.e., $V_g=0$ V). Since potential is relative, the potentials of portions of the semiconductor memory device are assumed, in the description below, to become 0 V after the interruption of power from the external power supply, although it may take slightly longer or shorter.

REFERENCES

- [Patent Document 1] U.S. Pat. No. 4,777,625
- [Patent Document 2] United States Patent Application Publication No. 2011/0101351
- [Patent Document 3] United States Patent Application Publication No. 2011/0156027
- [Patent Document 4] United States Patent Application Publication No. 2011/0182110

SUMMARY OF THE INVENTION

It is an object of one embodiment of the present invention to provide a memory device that is formed using a semiconductor with an irreversible conductivity type and a bandgap of 2.5 electron volts or more, such as an oxide semiconductor, is sufficiently integrated, and is capable of retaining data for a needed period even during interruption of power from an external power supply. It is another object of one embodiment of the present invention to provide a memory device with a novel structure or a method for driving the memory device, particularly a memory device whose power consumption can be reduced or a method for driving the memory device.

The terms used in this specification for the description of the present invention are briefly described. First, when one of a source and a drain of a transistor is called a drain, the other is called a source in this specification. That is, they are not distinguished depending on the potential level. Therefore, a portion called a source in this specification can be alternatively referred to as a drain.

Further, even when the expression "be connected" is used in this specification, there is a case in which no physical connection is made in an actual circuit and a wiring is just extended. For example, in a transistor circuit, there is a case in which one wiring serves as gates of a plurality of transistors. In that case, one wiring may have a plurality of branches to gates in a circuit diagram. In this specification, the expression "a wiring is connected to a gate" is also used to describe such a case.

One embodiment of the present invention is a semiconductor memory device, which includes a column driver, at least one bit line, at least one word line, at least one memory cell, and a transistor including a back gate (a back gate transistor). The memory cell includes a transistor and a capacitor. A source of the transistor is connected to the bit line. A drain of the transistor is connected to one electrode of the capacitor. A gate of the transistor is connected to the word line. A drain of the back gate transistor is connected to the bit line. A source of the back gate transistor is connected to the column driver. The potential of the back gate of the back gate transistor is lower than the minimum potential of the word line.

One embodiment of the present invention is a semiconductor memory device, which includes a column driver, at least one bit line, at least one word line, and at least one memory cell. The memory cell includes a transistor and a capacitor. A source of the transistor is connected to the bit line. A drain of the transistor is connected to one electrode of the capacitor. A gate of the transistor is connected to the word line. The bit line is connected to the column driver. The bit line includes a back gate transistor at one end. The potential of a back gate of the back gate transistor is lower than the minimum potential of the word line.

One embodiment of the present invention is a semiconductor memory device, which includes a column driver, at least one bit line, at least one word line, at least one memory cell, and a back gate transistor. The memory cell includes a transistor and a capacitor. A source of the transistor is connected to the bit line. A drain of the transistor is connected to one electrode of the capacitor. A gate of the transistor is connected to the word line. The bit line is connected to the column driver. The back gate transistor is inserted in series in the bit line. The potential of a back gate of the back gate transistor is lower than the minimum potential of the word line.

In the above-described semiconductor memory devices, two or more back gate transistors may be inserted in the bit line. A wiring connected to the back gate of the back gate transistor may be in a floating state. The back gate of the back gate transistor may be connected to a negative electrode of a battery which is additionally provided in the semiconductor memory device. Further, the drain of the transistor in a memory cell may be connected to a gate of another transistor in that memory cell. Furthermore, one or more sense amplifiers may be inserted in the bit line. The semiconductor memory device preferably includes a circuit for controlling a gate of the back gate transistor.

Moreover, it is preferable that the minimum potential of the bit line be higher than the minimum potential of the word line by 1 V or more. Further, it is preferable that the maximum potential of the gate of the back gate transistor be higher than the maximum potential of the word line by 1 V or more. Furthermore, it is preferable that the potential of the back gate of the back gate transistor be lower than the potential of any other portion.

Note that the foregoing applies to the case where the transistor in the memory cell and the back gate transistor are n-channel transistors. In the case where the transistor in the memory cell and the back gate transistor are p-channel transistors, the above-described potential combination is reversed; the one expressed with "high" and the one expressed with "low" in the above description are expressed with "low" and "high", respectively, and the one expressed with "maximum" and the one expressed with "minimum" in the above description are expressed with "minimum" and "maximum", respectively.

First, the effect of the back gate transistor will be described. The back gate transistor has a structure in which a semiconductor layer is sandwiched between the gate and the back gate. In one embodiment of the present invention, the potential of the back gate is preferably set constant. The off-state current of the back gate transistor can be sufficiently reduced by using a semiconductor with a wide bandgap as described above.

The drain current I_d of an n-channel transistor which does not have a back gate is represented by a curve A in FIG. 5A. It is assumed here that the source potential V_s of the transistor is 0 V and the drain potential V_d thereof is higher than 0 V. As illustrated, at a gate potential V_g of 0 V, the drain current I_d is considerably large. On the other hand, at a gate potential V_g of $-V_1$ (<0), the drain current I_d is at a negligible level. The value of V_1 may be set as appropriate depending on the structure of the transistor or the like, but is preferably higher than or equal to +1 V.

Note that the minimum of the drain current I_d ideally depends on the bandgap of a semiconductor, and that of a transistor formed using a semiconductor with a bandgap of 3.2 electron volts and with no defects (with the channel length and the channel width equal to each other and with short channel effects not taken into consideration), for example, is approximately 10^{-31} A.

On the other hand, when the potential of the back gate of the back gate transistor is set to an appropriate value, the drain current I_d can be sufficiently small even at a gate potential V_g of 0 V. For example, when the potential of the back gate is set so that the potential of a gate-side surface of the semiconductor layer of the transistor is substantially equal to or lower than $-V_1$ at a gate potential V_g of 0 V, the drain current I_d is represented by a curve B in FIG. 5A. In other words, at the gate potential V_g of 0 V, the drain current I_d is sufficiently small and is at a negligible level.

This is largely attributed to the suppression of leakage current (which is due to a short channel effects) on the back side (the side opposite to the gate) of the semiconductor layer which is achieved with the back gate having a negative potential, and also to the resulting decrease in the subthreshold value. Note that the threshold voltage can also be largely changed according to the potential of the back gate.

The back gate is preferably held at a constant potential, and for that purpose, the back gate may be in a floating state. For example, the back gate may be connected to one electrode of a capacitor which is provided to retain electric charge of the back gate. Alternatively, the back gate may be connected to a negative electrode of a battery which is additionally provided in the semiconductor memory device. In any case, the amount of charge released from the back gate to the outside is significantly small, and a potential change in the capacitor and a battery drain are quite limited.

The use of the back gate transistor as described above enables the drain current to be sufficiently small even in the state where power is not supplied from the outside (the state where both the gate potential and the source potential are 0 V). However, in some cases, it may be difficult to use such a back gate transistor as a transistor in every memory cell.

In terms of the structure of the back gate transistor, it is necessary to add a back gate to an ordinary transistor, which may cause an increase in the number of processes. In addition, since the back gate is provided, the degree of integration in a circuit design may be lowered. Furthermore, in the case where there is a large potential difference between the back gate and another circuit, the back gate needs to be located sufficiently away from the circuit, which may also cause the degree of integration to be lowered.

To solve such problems, the present inventor has found that all memory cells can have sufficient retention characteristics by inserting a small number of back gate transistors in appropriate portions of bit lines.

In one embodiment of the present invention described above, the back gate transistor is provided, for example, between the column driver and the bit line to put the bit line in a floating state; thus, the potential of the bit line can be kept constant. If the potential of the bit line is a constant value, the drain current can be sufficiently reduced even when the gate of the transistor in the memory cell has a potential of 0 V. This is described with reference to FIGS. 5A to 5C.

FIG. 5B illustrates a back gate transistor 108 inserted in a bit line 104. A drain of the back gate transistor 108 is connected to the bit line 104, and a source thereof is connected to a column driver 102. A gate of the back gate transistor 108 is connected to a bit line controlling line 112, and a back gate thereof is connected to a back gate line 111. Note that the back gate line 111 is constantly held at a potential V_3 (<0 V).

A bit line capacitance 121 exists in the bit line 104. Most of the bit line capacitance 121 is a parasitic capacitance and is usually 10 fF or more, typically 100 fF or more, although it depends on the length of the bit line 104, the circuit configuration, or the like. It is needless to say that a capacitance intentionally provided in parallel to the bit line may be used as part of the bit line capacitance 121.

The potential of the bit line 104, which changes according to written or read data, is set to the value V_1 (>0 V) or more while the semiconductor memory device is powered by the external power supply. It is assumed here that the potential of the bit line 104 is V_1 . While the semiconductor memory device is powered by the external power supply, the potential of the bit line controlling line 112 is an appropriate

positive value (e.g., V_2 (>0 V)), so that the back gate transistor 108 is in the on state as represented by the curve B in FIG. 5A.

If power from the external power supply is interrupted here, the semiconductor memory device detects the interruption of power and sets the potential of the bit line controlling line 112 to 0 V or less. Consequently, the back gate transistor 108 is turned off. When the potentials of many portions of the semiconductor memory device sufficiently decrease, the potential of the bit line controlling line 112 becomes 0 V. In addition, the potential of the column driver 102 also become 0 V; thus, the source potential of the back gate transistor 108 also becomes 0 V.

However, the back gate transistor 108 is turned off at the same time as interruption of power from the external power supply, so that the potential of the bit line 104 (the drain of the back gate transistor 108) remains at V_1 . Furthermore, the drain current of the back gate transistor 108 at a gate potential of 0 V is significantly small as represented by the curve B in FIG. 5A; thus, the potential of the bit line 104 can be retained at a value close to V_1 for a very long period of time.

The bit line 104 is connected to the memory cell 105. While the semiconductor memory device is powered by the external power supply, the potential of the drain of the cell transistor 106 in the memory cell 105, which changes according to written data as shown in FIG. 5C, is V_1 or more because the potential of the bit line 104 is V_1 or more. It is assumed here that the potential of the drain of the cell transistor 106 is V_4 ($\geq V_1$).

After interruption of power from the external power supply, the potential of the bit line 104 is V_1 as described above, so that the potential of the source of the cell transistor is V_1 . On the other hand, the potential of the word line 103 (the potential of the gate of the cell transistor 106) becomes 0 V owing to interruption of power from the external power supply. The drain current of the cell transistor in this state is equivalent to that in the case where the gate potential V_g is set to $-V_1$ in the curve A of FIG. 5A. In other words, the drain current becomes a very low value and charge in the capacitor 107 can be retained for a sufficient period of time.

That is, by inserting the back gate transistor 108 in the bit line 104, the potential of the bit line 104 can be retained at an appropriate positive value for a sufficient period of time even during interruption of power from the external power supply. Thus, the semiconductor memory device can exhibit sufficient data retention characteristics even when using cell transistors with a variety of channel lengths and threshold voltages at a wider range of temperatures. Placing a limited number of back gate transistors can produce an effect equivalent to the case of using a back gate transistor in every memory cell.

Since the back gate transistor 108 is inserted in series in the bit line 104, its resistance in the on state is desirably as low as possible. An effective way to accomplish this is to increase the potential of the gate of the back gate transistor 108. For example, the potential of the gate of the back gate transistor 108 is preferably higher than the maximum potential of the gate of another transistor (e.g., the maximum potential of the word line 103) by 1 V or more. Alternatively, the channel width of the back gate transistor may be 10 or more times as large as the minimum feature size.

By setting the minimum potential of the word line 103 to $-V_1$ in the state where the semiconductor memory device is powered by the external power supply, the resistance of the

cell transistor **106** in the off state can be sufficiently increased and charge accumulated in the capacitor **107** can be retained.

BRIEF DESCRIPTION OF THE DRAWINGS

FIGS. **1A** and **1B** each illustrate an example of a semiconductor memory device according to the present invention.

FIGS. **2A** and **2B** illustrate an example of a conventional semiconductor memory device.

FIG. **3** illustrates an example of a semiconductor memory device according to the present invention.

FIGS. **4A** and **4B** illustrate an example of a semiconductor memory device according to the present invention.

FIGS. **5A** to **5C** illustrate principles of one embodiment of the present invention.

FIGS. **6A** to **6C** illustrate an example of a manufacturing process of a semiconductor memory device according to the present invention.

FIGS. **7A** and **7B** illustrate an example of a manufacturing process of a semiconductor memory device according to the present invention.

FIG. **8** illustrates an example of a manufacturing process of a semiconductor memory device according to the present invention.

FIGS. **9A** to **9D** illustrate an example of a manufacturing process of a semiconductor memory device according to the present invention.

FIGS. **10A** to **10C** illustrate an example of a manufacturing process of a semiconductor memory device according to the present invention.

DETAILED DESCRIPTION OF THE INVENTION

Hereinafter, embodiments will be described with reference to drawings. However, the embodiments can be implemented with various modes. It will be readily appreciated by those skilled in the art that modes and details can be changed in various ways without departing from the spirit and scope of the present invention. Thus, the present invention should not be interpreted as being limited to the following description of the embodiments.

Embodiment 1

In this embodiment, a semiconductor memory device in FIG. **1A** will be described. The semiconductor memory device in FIG. **1A** includes a row driver **101**, a column driver **102**, a plurality of word lines **103** connected to the row driver **101**, a plurality of bit lines **104** (indirectly) connected to the column driver **102**, and memory cells **105** each provided at the intersection of the word line **103** and the bit line **104**. This structure is the same as that of the conventional DRAM in FIGS. **2A** and **2B**.

The semiconductor memory device in FIG. **1A** further includes back gate transistors **108** each inserted between the column driver **102** and the bit line **104**. It can also be said that the back gate transistor **108** is inserted in the bit line **104**. It can also be said that a source of the back gate transistor **108** is connected to the column driver **102**, and a drain of the back gate transistor **108** is connected to the bit line **104**. It can also be said that the back gate transistor **108** is inserted between the column driver **102** and the memory cell **105** that is the closest to the column driver **102**.

A gate of the back gate transistor **108** is connected to a bit line controlling line **112**, and a back gate of the back gate transistor **108** is connected to a back gate line **111**. The potential of the bit line controlling line **112** is set by a bit line controlling circuit **110**. The back gate line **111** is connected to one electrode of a capacitor **109**, and the potential thereof is held at an appropriate negative value regardless of whether or not an external power supply is provided.

For that purpose, charge may be injected so that the potential of the capacitor **109** (the back gate line **111**) becomes appropriate, and then the back gate line **111** may be brought into a floating state. Alternatively, with the back gate line **111** placed in a floating state, an electron beam with an energy of several tens of kilo electron volts or higher may be injected into part thereof.

Note that as illustrated in FIG. **1B**, the back gate line **111** may be connected to a negative electrode of a battery **113** provided over a substrate where the semiconductor memory device is formed or in a package including the substrate. Since the amount of current flowing through the back gate line **111** is significantly small, the capacity of the battery **113** can be extremely low.

When the semiconductor memory device is powered by an external power supply and is determined to be usable, the bit line controlling circuit **110** supplies the bit line controlling line **112** with an appropriate potential to turn on the back gate transistor **108**. When interruption of power from the external power supply is detected, or termination of the use of the semiconductor memory device is detected even while the semiconductor memory device is powered by the external power supply, the bit line controlling circuit **110** sets the potential of the bit line controlling line **112** to 0 V or less to rapidly turn off the back gate transistor **108**.

Embodiment 2

A semiconductor memory device according to this embodiment will be described with reference to FIG. **3**. The semiconductor memory device in FIG. **3** has sense amplifiers **114** inserted in bit lines **104**. The sense amplifier **114** is used to divide the bit line **104** into appropriate lengths to lower the bit line capacitance during read operation so that read accuracy can be increased.

With the sense amplifier **114** inserted in the bit line **104** in this manner, when power from an external power supply is interrupted, for example, the charge in the bit line **104** flows out also through the sense amplifier **114**. As a result, when power from the external power supply is interrupted, the potential of the bit line **104** connected to the sense amplifier decreases to 0 V.

Therefore, it is necessary to prevent charge in the bit lines **104** from flowing out when power from the external power supply is interrupted, by providing the back gate transistors such that the sense amplifier **114** is sandwiched therebetween.

The semiconductor memory device in FIG. **3** includes a column driver **102**, a plurality of word lines **103**, a plurality of bit lines **104**, and memory cells **105** each provided at the intersection of the word line **103** and the bit line **104**. In addition, the sense amplifier **114** is inserted in the bit line **104**.

Like the semiconductor memory device described in Embodiment 1, the semiconductor memory device in FIG. **3** further includes back gate transistors **108_1** each inserted between the column driver **102** and the bit line **104**. A gate of the back gate transistor **108_1** is connected to a bit line controlling line **112_1**, and a back gate of the back gate

transistor **108_1** is connected to a back gate line **111_1**. The potential of the back gate line **111_1** is held at an appropriate negative value regardless of whether or not the external power supply power is provided.

The semiconductor memory device in FIG. 3 further includes back gate transistors **108_2** and back gate transistors **108_3** each provided between the sense amplifier **114** connected to the bit line **104** and the memory cell **105** that is the closest to the sense amplifier **114**. A gate of the back gate transistor **108_2** is connected to a bit line controlling line **112_2**, and a gate of the back gate transistor **108_3** is connected to a bit line controlling line **112_3**. A back gate of the back gate transistor **108_2** is connected to a back gate line **111_2**, and a back gate of the back gate transistor **108_3** is connected to a back gate line **111_3**. The potentials of the back gate line **111_2** and the back gate line **111_3** are each held at an appropriate negative value regardless of whether or not the external power supply power is provided.

In such a semiconductor memory device, the potentials of the bit line controlling lines **112_1** to **112_3** change according to conditions in a manner similar to that in Embodiment 1. In other words, when the semiconductor memory device is powered by an external power supply and is usable, the bit line controlling lines **112_1** to **112_3** are supplied with such a potential that the back gate transistors **108_1** to **108_3** are turned on.

In contrast, when power from the external power supply is interrupted or when the use of the semiconductor memory device is terminated even while the semiconductor memory device is powered by the external power supply, the bit line controlling lines **112_1** to **112_3** are supplied with such a potential that the back gate transistors **108_1** to **108_3** are turned off.

For example, when power from the external power supply is interrupted, the potential of the bit line controlling lines **112_1** to **112_3** rapidly becomes 0 V or less to turn off the back gate transistors **108_1** to **108_3**. Consequently, the bit line **104** is divided by the back gate transistors **108_1** to **108_3**. Thus, even if the potential of portions of the bit line **104** connected to the column driver **102** and the sense amplifier **114** becomes 0 V, the potential of the other portions (portions connected to the memory cells **105**) can remain at an appropriate value (>0 V).

On the other hand, because the potential of the word line **103** is 0 V, the cell transistor in the memory cell has sufficiently high resistance, and thus enables charge accumulated in the capacitor to be retained for a long period of time.

Embodiment 3

A semiconductor memory device in FIGS. 4A and 4B will be described. Memory cells **117** in the semiconductor memory device in FIGS. 4A and 4B have the same configuration as those described in Patent Document 4. Refer to Patent Document 4 for the operation and the like of the memory cells **117**.

As illustrated in FIG. 4B, the memory cell **117** according to this embodiment includes a write transistor **118**, a read transistor **119**, and a capacitor **120**. A source of the write transistor **118** and a source of the read transistor **119** are connected to a bit line **104**. A gate of the write transistor **118** is connected to a write word line **115**. A drain of the write transistor **118** and a gate of the read transistor **119** are connected to one electrode of the capacitor **120**. The other electrode of the capacitor **120** is connected to a read word line **116**.

The potentials of the write word lines **115** and the read word lines **116** are controlled by a row driver **101**. The potential of the bit line **104** is controlled by a column driver **102**.

While there are such many differences between the memory cell **105** according to Embodiment 1 or 2 and the memory cell **117** according to this embodiment, these memory cells are the same in that the source of the write transistor **118** (which corresponds to the cell transistor **106** in the memory cell **105** in FIG. 1A) is connected to the bit line **104** and the drain of the write transistor **118** is connected to one electrode of the capacitor **120**. In other words, for data retention, the write transistor **118** needs to exhibit high resistance in the off state.

Thus, in a manner similar to that in Embodiments 1 and 2, back gate transistors **108** each inserted between the column driver **102** and the bit line **104** are provided, which can achieve sufficiently high resistance even when power from an external power supply is interrupted (see FIG. 4A). A gate of the back gate transistor **108** is connected to a bit line controlling line **112**, and a back gate of the back gate transistor **108** is connected to a back gate line **111**. The potential of the back gate line **111** is held at an appropriate negative value regardless of whether or not the external power supply is provided.

In such a semiconductor memory device, the potential of the bit line controlling line **112** changes according to conditions in a manner similar to that in Embodiment 1. In other words, when the semiconductor memory device is powered by the external power supply and is usable, the bit line controlling line **112** is supplied with such a potential that the back gate transistor **108** is turned on.

In contrast, when power from the external power supply is interrupted or when the use of the semiconductor memory device is terminated even while the semiconductor memory device is powered by the external power supply, the bit line controlling line **112** is supplied with such a potential that the back gate transistor **108** is turned off.

For example, when power from the external power supply is interrupted, the potential of the bit line controlling line **112** rapidly becomes 0 V or less to turn off the back gate transistor **108**. Consequently, the potential of the bit line **104** can remain at an appropriate value (>0 V).

On the other hand, because the potential of the write word line **115** is 0 V, the write transistor **118** in the memory cell **117** has sufficiently high resistance, and thus enables charge in the capacitor **120** to be retained for a long period of time.

The memory cell **117** is characterized by being capable of amplifying a signal with the read transistor **119** and output the amplified signal to the bit line even if the capacitance of the capacitor **120** is low. However, the fact that the capacitance of the capacitor **120** is low means that it is difficult to retain data for a required time if the resistance of the write transistor **118** in the off state is not sufficiently high. Therefore, keeping, during power interruption, the potential of the bit line **104** at an appropriate positive value with the back gate transistor **108** to increase the resistance of the write transistor **118** in the off state is particularly effective in this embodiment.

Embodiment 4

A brief description is given of a process for manufacturing the semiconductor memory device illustrated in, for example, FIGS. 1A and 1B or FIG. 3 with reference to FIGS. 6A to 6C, FIGS. 7A and 7B, and FIG. 8. Refer to known semiconductor integrated circuit manufacturing techniques

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for the details. Note that FIGS. 6A to 6C, FIGS. 7A and 7B, and FIG. 8 illustrate the concepts of the manufacturing process and do not show specific cross sections.

<FIG. 6A>

First, device isolation insulators **202**, n-type impurity regions **203N**, p-type impurity regions **203P**, an n-channel transistor gate **204N**, a p-channel transistor gate **204P**, a first interlayer insulator **205**, first contact plugs **206a** to **206d**, and the like are formed over a surface of a substrate **201** of a semiconductor or the like by known semiconductor integrated circuit manufacturing techniques. The n-channel transistor or the p-channel transistor here may be used in a row driver, a column driver, a sense amplifier, or the like in a semiconductor memory device.

<FIG. 6B>

Next, first layer wirings **208a** to **208d** are formed so as to be embedded in a first embedment insulator **207**. These wirings are used in, for example, the row driver **101** or the column driver **102** in FIGS. 1A and 1B, or the sense amplifier **114**.

<FIG. 6C>

Further, a second interlayer insulator **209**, a second contact plug **210**, a second embedment insulator **211**, second layer wirings **212a** to **212c** are formed. Here, the second layer wiring **212b** corresponds to the back gate of the back gate transistor **108** or the back gate line **111** in FIGS. 1A and 1B. Note that one or more layers including another wiring may be additionally provided between a layer including the second layer wirings **212a** to **212c** and a layer including the first layer wirings **208a** to **208d**.

<FIG. 7A>

Further, a third interlayer insulator **213**, third contact plugs **214a** to **214c**, a third embedment insulator **215**, and third layer wirings **216a** to **216e** are formed. Note that the first contact plug **206a**, the first layer wiring **208a**, the second contact plug **210**, the second layer wiring **212a**, the second layer wiring **212c**, the third contact plug **214a**, the third contact plug **214b**, the third layer wiring **216a**, and the third layer wiring **216b** serve as part of the bit line **104** in FIGS. 1A and 1B.

<FIG. 7B>

Subsequently, an oxide semiconductor layer **217a** and an oxide semiconductor layer **217b** are formed, and a gate insulator **218** is formed so as to cover them. At this time, it is preferable that the physical thickness of the gate insulator **218** be two or more times that of the oxide semiconductor layer **217a** and the oxide semiconductor layer **217b**, because this enables the oxide semiconductor layer **217a** and the oxide semiconductor layer **217b** to be covered with the gate insulator **218** reliably, thereby preventing shorts between wirings.

On the other hand, it is preferable that the effective thickness of the gate insulator (e.g., the equivalent oxide thickness) be less than or equal to that of the oxide semiconductor layer **217a** and the oxide semiconductor layer **217b**. Therefore, it is preferable that the gate insulator **218** be formed using a material whose dielectric constant is twice that of the oxide semiconductor layer **217a** and the oxide semiconductor layer **217b**.

For example, the gate insulator **218** may be formed using a high dielectric constant material such as hafnium oxide, tantalum oxide, or zirconium oxide. Materials such as barium oxide, strontium oxide, calcium oxide, and lithium oxide which form silicides on silicon semiconductor have been prevented from being used with silicon semiconductor, but may be used with an oxide semiconductor without

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problems. Therefore, any of these materials can be used for the gate insulator **218** as long as it has high dielectric constant.

Then, fourth layer wirings **219a** to **219d** are formed. The fourth layer wiring **219a** here corresponds to the gate of the back gate transistor **108** or the bit line controlling line **112** in FIG. 1A. The fourth layer wirings **219b** to **219d** correspond to the word lines **103** in FIG. 1A.

<FIG. 8>

Stacked capacitors are formed by known DRAM manufacturing techniques. Specifically, a fourth interlayer insulator **220**, a fourth contact plug **221a**, and a fourth contact plug **221b** are formed, and then a fifth interlayer insulator **222**, a capacitor electrode **223a** and a capacitor electrode **223b** are formed thereover. Subsequently, a capacitor insulator **224** and a cell plate **225** are formed. Thus, the semiconductor memory device can be manufactured.

Embodiment 5

A brief description is given of a process for manufacturing the semiconductor memory device illustrated in FIGS. 4A and 4B with reference to FIGS. 9A to 9D and FIGS. 10A to 10C. Refer to known semiconductor integrated circuit manufacturing techniques or Patent Document 2 for the details. Note that FIGS. 9A to 9D and FIGS. 10A to 10C illustrate the concepts of the manufacturing process and do not show specific cross sections.

<FIG. 9A>

First, a BOX layer **302**, an SOI layer **303a**, and an SOI layer **303b** are formed over a surface of a substrate **301** of a semiconductor or the like by known semiconductor integrated circuit manufacturing techniques.

<FIG. 9B>

Next, read gates **304a** and **304b** are formed, and an impurity is added to the SOI layer **303a** and the SOI layer **303b** by using these gates as a mask to form impurity regions **305a** to **305d**. Here, the impurity region **305a** corresponds to the back gate of the back gate transistor **108** or the back gate line **111** in FIG. 4A. The read gates **304a** and **304b** correspond to the gates of the read transistors **119** in FIGS. 4A and 4B. Then, a first interlayer insulator **306** is formed and then is planarized to expose top surfaces of the read gates **304a** and **304b**.

<FIG. 9C>

First layer wirings **307a** to **307e** and a first embedment insulator **308** are formed.

<FIG. 9D>

Subsequently, an oxide semiconductor layer **309a** and an oxide semiconductor layer **309b** are formed, and a gate insulator **310** is formed so as to cover them. Then, second layer wirings **311a** to **311e** are formed. The second layer wiring **311a** here corresponds to the gate of the back gate transistor **108** or the bit line controlling line **112** in FIG. 4A. The second layer wirings **311c** and **311d** correspond to the write word lines **115** in FIGS. 4A and 4B. The second layer wirings **311b** and **311e** correspond to the read word lines **116** in FIGS. 4A and 4B.

<FIG. 10A>

A second interlayer insulator **312** with a plane surface is formed. Then, contact plugs **313a**, **313b**, and **313c** connected to the first layer wirings **307a**, **307b**, and **307d** are formed.

<FIG. 10B>

Third layer wirings **314a** and **314b** are formed. The third layer wirings **314a** and **314b** correspond to the bit line **104** in FIG. 4A.

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<FIG. 10C>

A third interlayer insulator **315** is formed. Any other wirings, interlayer insulators, and the like may additionally be formed. Through the aforementioned process, a semiconductor memory device including a back gate transistor **316**, a read transistor **317**, a write transistor **318**, and a capacitor **319** is formed. The back gate transistor **316** corresponds to the back gate transistor **108** in FIG. 4A.

The read transistor **317**, the write transistor **318**, and the capacitor **319** form one memory cell. The read transistor **317**, the write transistor **318**, and the capacitor **319** correspond to the read transistor **119**, the write transistor **118**, and the capacitor **120** in FIG. 4B, respectively.

Note that FIG. 10C illustrates two memory cells (a memory cell **320a** and a memory cell **320b**). These memory cells are connected to the same bit line.

This application is based on Japanese Patent Application serial no. 2011-129685 filed with Japan Patent Office on Jun. 10, 2011, the entire contents of which are hereby incorporated by reference.

What is claimed is:

1. A semiconductor memory device comprising:
 - a column driver;
 - a bit line;
 - a word line;
 - a memory cell comprising a first transistor and a capacitor; and
 - a second transistor including a back gate,
 - wherein a source of the first transistor is electrically connected to the bit line,
 - wherein a drain of the first transistor is electrically connected to one electrode of the capacitor,
 - wherein a gate of the first transistor is electrically connected to the word line,
 - wherein a drain of the second transistor is electrically connected to the bit line,
 - wherein a source of the second transistor is electrically connected to the column driver,
 - wherein a potential of the back gate of the second transistor is lower than a minimum potential of the word line, and
 - wherein each of the first transistor and the second transistor comprises a semiconductor with a bandgap of 2.5 electron volts or more.
2. The semiconductor memory device according to claim 1, further comprising a wiring electrically connected to the back gate of the second transistor,
 - wherein the wiring is in a floating state.
3. The semiconductor memory device according to claim 1,
 - wherein the memory cell further comprises a third transistor, and
 - wherein the drain of the first transistor is electrically connected to a gate of the third transistor.
4. The semiconductor memory device according to claim 1, wherein a minimum potential of the bit line is higher than the minimum potential of the word line by 1 V or more.
5. The semiconductor memory device according to claim 1, further comprising one or more sense amplifiers inserted in the bit line.
6. The semiconductor memory device according to claim 1, wherein a maximum potential of a gate of the second transistor is higher than a maximum potential of the word line by 1 V or more.
7. The semiconductor memory device according to claim 1, wherein the potential of the back gate of the second transistor is lower than a potential of any other portion.

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8. The semiconductor memory device according to claim 1, wherein each of the first transistor and the second transistor comprises a semiconductor with an irreversible conductivity type.

9. The semiconductor memory device according to claim 1, wherein each of the first transistor and the second transistor comprises an oxide semiconductor.

10. The semiconductor memory device according to claim 1, wherein the back gate of the second transistor is electrically connected to a negative electrode of a battery.

11. A semiconductor memory device comprising:

- a column driver;
- a bit line;
- a word line;
- a memory cell comprising a first transistor and a capacitor; and
- a second transistor including a back gate,
 - wherein a source of the first transistor is electrically connected to the bit line,
 - wherein a drain of the first transistor is electrically connected to one electrode of the capacitor,
 - wherein a gate of the first transistor is electrically connected to the word line,
 - wherein the bit line is electrically connected to the column driver,
 - wherein the second transistor is inserted in series in the bit line,
 - wherein a potential of the back gate of the second transistor is lower than a minimum potential of the word line, and
 - wherein each of the first transistor and the second transistor comprises a semiconductor with a bandgap of 2.5 electron volts or more.

12. The semiconductor memory device according to claim 11, further comprising a wiring electrically connected to the back gate of the second transistor,

- wherein the wiring is in a floating state.

13. The semiconductor memory device according to claim 11,

- wherein the memory cell further comprises a third transistor, and
- wherein the drain of the first transistor is electrically connected to a gate of the third transistor.

14. The semiconductor memory device according to claim 11, wherein a minimum potential of the bit line is higher than the minimum potential of the word line by 1 V or more.

15. The semiconductor memory device according to claim 11, further comprising one or more sense amplifiers inserted in the bit line.

16. The semiconductor memory device according to claim 11, wherein a maximum potential of a gate of the second transistor is higher than a maximum potential of the word line by 1 V or more.

17. The semiconductor memory device according to claim 11, wherein the potential of the back gate of the second transistor is lower than a potential of any other portion.

18. The semiconductor memory device according to claim 11, wherein each of the first transistor and the second transistor comprises a semiconductor with an irreversible conductivity type.

19. The semiconductor memory device according to claim 11, wherein each of the first transistor and the second transistor comprises an oxide semiconductor.

20. The semiconductor memory device according to claim 11, wherein the back gate of the second transistor is electrically connected to a negative electrode of a battery.